

FM – Generation and Detection

Reference

- Chapter 5.3, Carlson, communication Systems.
- NBFM
- WBFM

Generation of FM signals

NBFM (Narrowband FM)

- NBFM can directly be implemented based on its general equation.

Carrier frequency

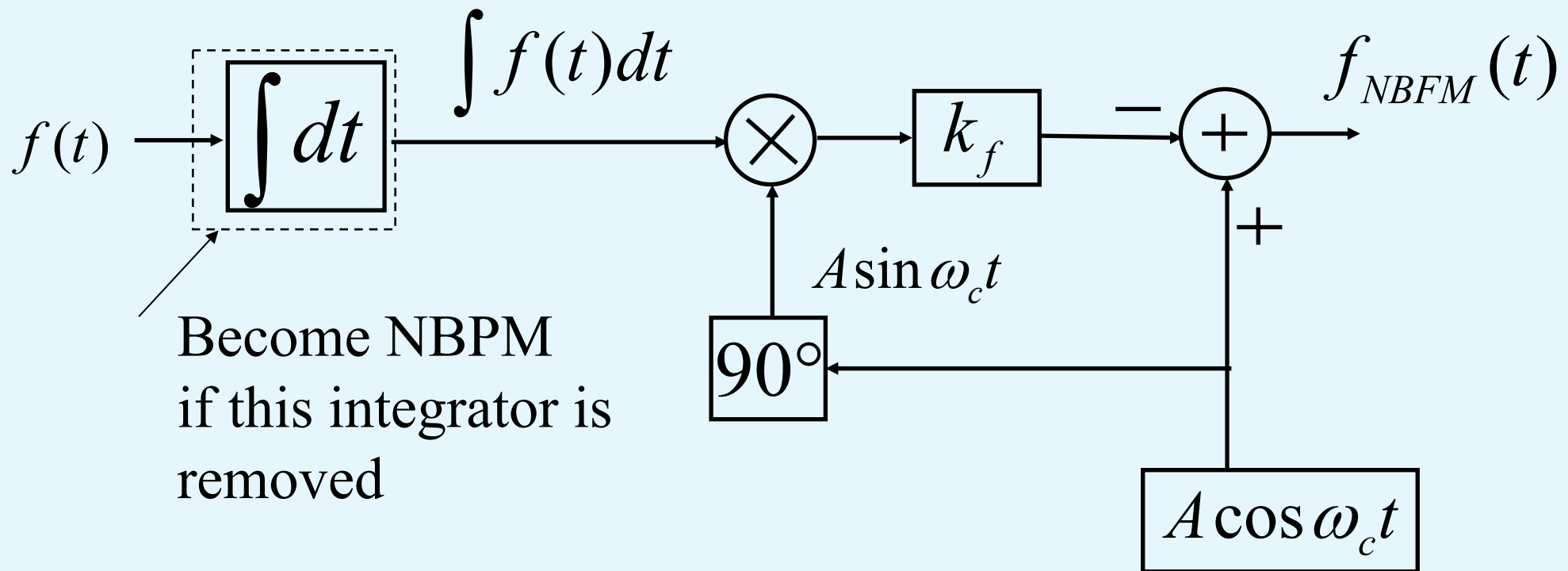
$$\begin{aligned} f_{NBFM}(t) &= A \cos \left[\omega_c t + k_f \int_0^t f(\tau) d\tau \right] \\ &= A \cos \omega_c t \cos \left[k_f \int_0^t f(\tau) d\tau \right] - A \sin \omega_c t \sin \left[k_f \int_0^t f(\tau) d\tau \right] \\ &\approx A \cos \omega_c t - A \sin \omega_c t \cdot \left[k_f \int_0^t f(\tau) d\tau \right] \quad \because \left| k_f \int_0^t f(\tau) d\tau \right| \ll 1 \end{aligned}$$

$$\cos x \approx 1 \quad \text{if } x \ll 1$$

$$\sin x \approx x \quad \text{if } x \ll 1$$

$$\cos(A + B) = \cos A \cos B - \sin A \sin B$$

NBFM

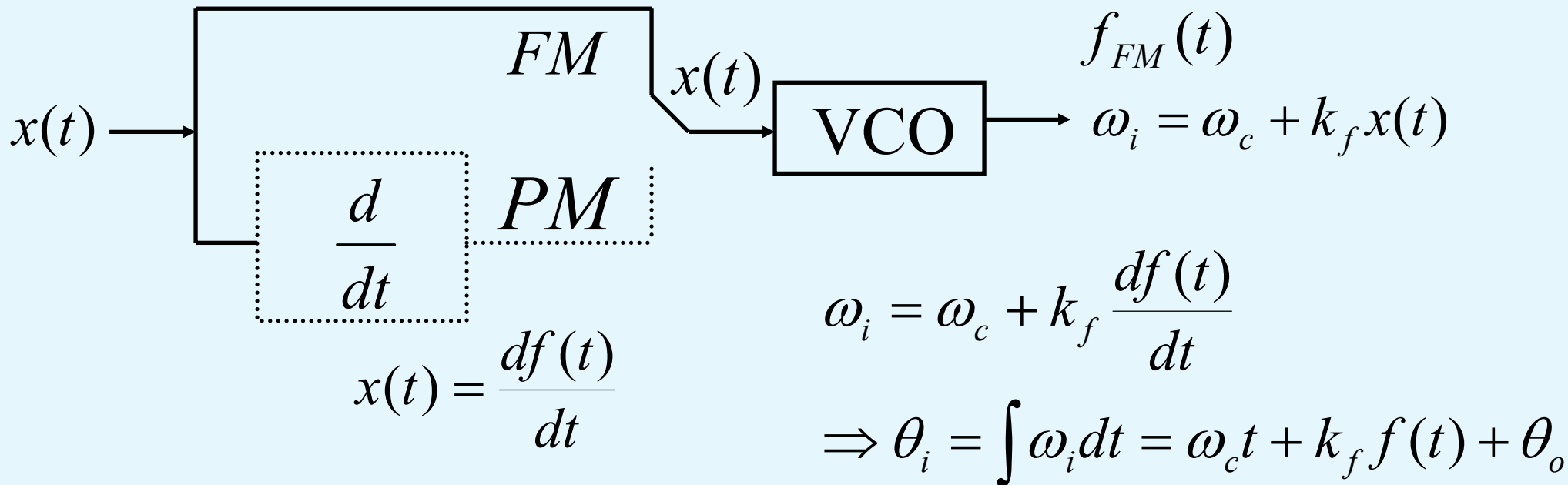


$$f_{PM} = A \cos[\omega_c t + k_p f(t)]$$

WBFM

A. Direct Method

- FM: $f_i \propto x(t)$
- Voltage-controlled oscillator (VCO)
 - $f_{out} \propto v_{in}$



WBFM

B. Indirect Method

- consists of a NBFM modulator, frequency multipliers and frequency converters.
- frequency multiplier
 - nonlinear device
 - multiply the frequencies of the input signal by a given factor
 - Example: an ideal square-law device

$$e_o(t) = ae_i^2(t)$$



WBFM

$$e_o(t) = ae_i^2(t)$$



If the input signal is a FM signal,

$$e_i(t) = A \cos(\omega_c t + \beta \sin \omega_m t)$$

the output signal is

$$e_o(t) = aA^2 \cos^2(\omega_c t + \beta \sin \omega_m t)$$

$$= \frac{1}{2} aA^2 [1 + \cos(2\omega_c t + 2\beta \sin \omega_m t)]$$

Can be removed
with a filter

A FM signal with carrier

frequency $2\omega_c$ and modulation index 2β

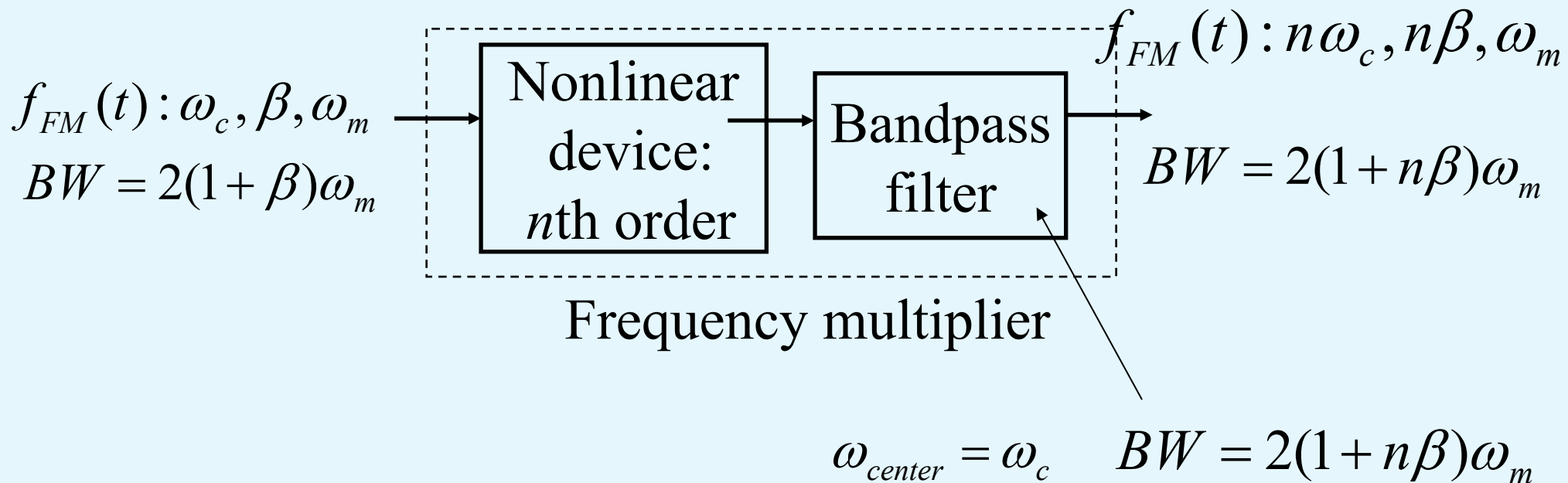
Modulating frequency
is unchanged

WBFM

n^{th} law device followed by a filter

– new FM signal

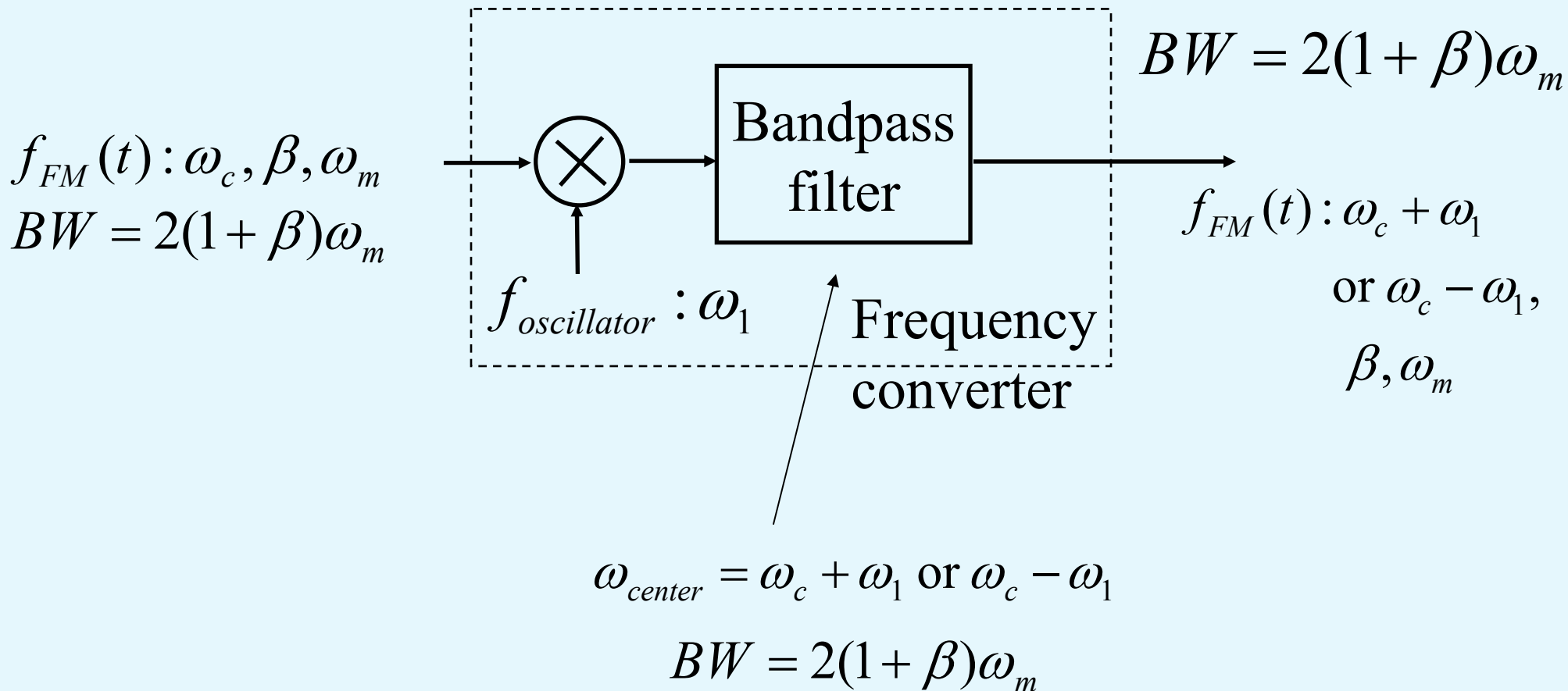
- carrier and a modulation index that have been increased by a factor of n .



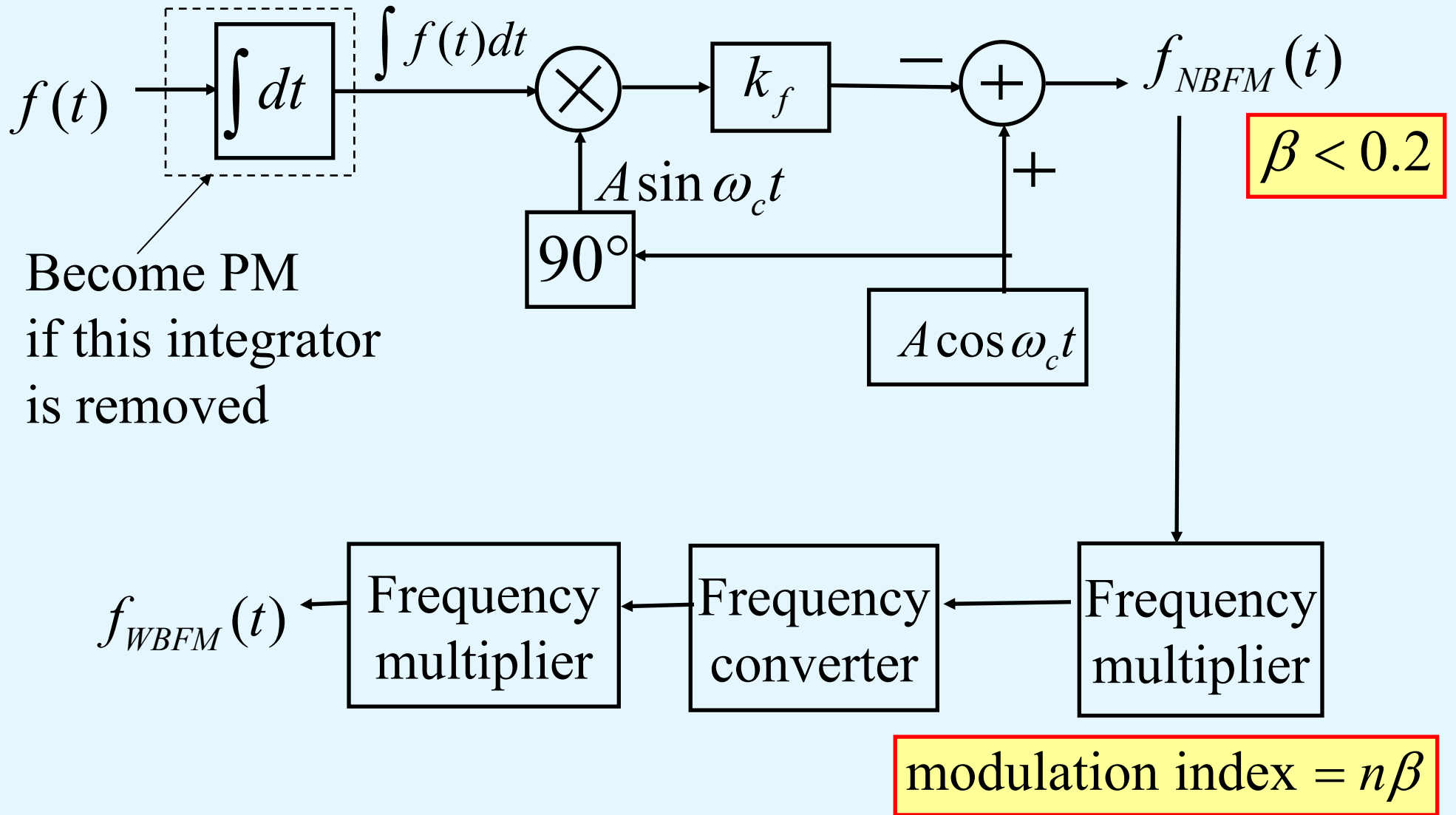
WBFM

Frequency converter

- to control the value of the carrier frequency.



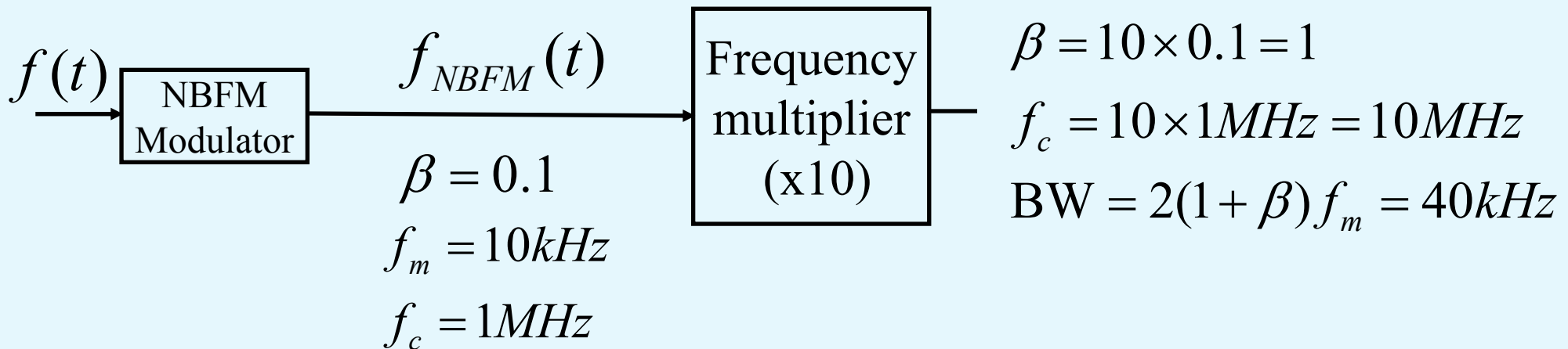
WBFM



WBFM

Example:

generate a FM signal with $f_c = 50\text{MHz}$ and $\beta = 1$ from a NBFM signal with $f_c = 1\text{MHz}$, $\beta = 0.1$ and $f_m = 10\text{kHz}$



WBFM

$$\beta = 10 \times 0.1 = 1$$

$$f_c = 10 \times 1\text{MHz} = 10\text{MHz}$$

$$\text{BW} = 2(1 + \beta)f_m = 40\text{kHz}$$

BPF

$$f_{center} = 10\text{MHz}$$

$$\text{BW} = 40\text{kHz}$$

Frequency
converter

$$f_{LO} = 40\text{MHz}$$

$$f_{WBFM}(t)$$

BPF

$$f_{center} = 50\text{MHz}$$

$$\text{BW} = 40\text{kHz}$$

$$\begin{aligned} f_{c(new)} &= f_{LO} \pm f_c \\ &= 40\text{MHz} \pm 10\text{MHz} \\ &= 50\text{MHz and } 30\text{MHz} \end{aligned}$$

$$\text{BW} = 40\text{kHz}$$

Demodulation of FM signals

Direct Method

- FM signal:

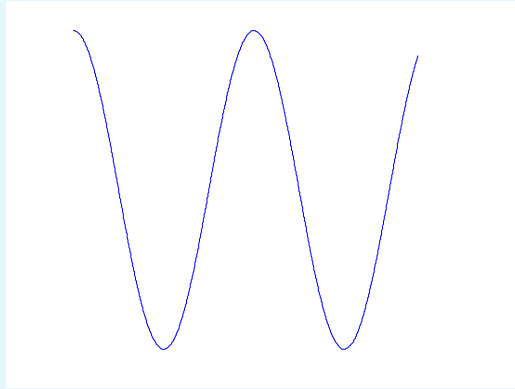
$$f_{FM}(t) = A \cos \left[\omega_c t + k_f \int_0^t f(\tau) d\tau \right]$$

$$\frac{d}{dt} f_{FM}(t) = -A \left[\omega_c + k_f f(t) \right] \cos \left[\omega_c t + k_f \int_0^t f(\tau) d\tau \right]$$

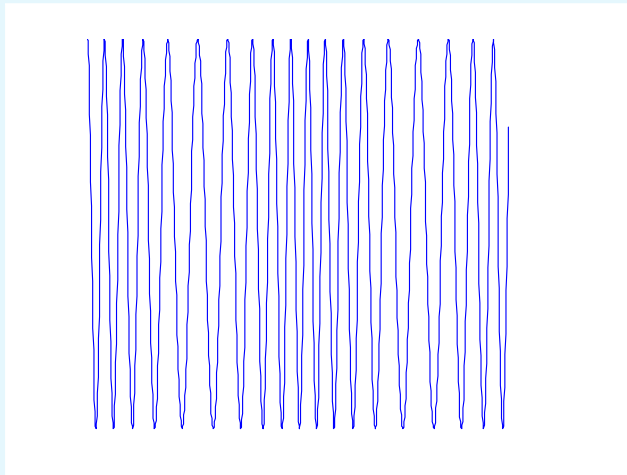
- If $k_f f(t) \ll \omega_c$, the equation is in the form of an AM signal whose envelope is

$$\approx A \omega_c \left[1 + \frac{k_f}{\omega_c} f(t) \right]$$

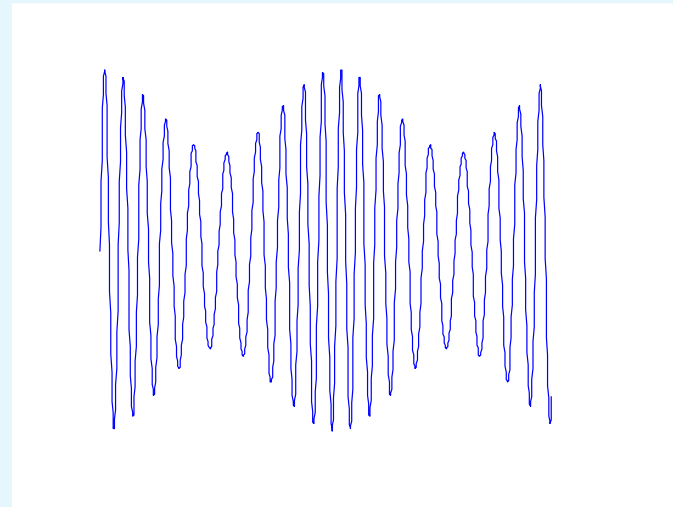
Demodulation of FM signals



$$\cos \omega_m t$$



$$\cos[10\omega_m t + 3 \sin \omega_m t]$$

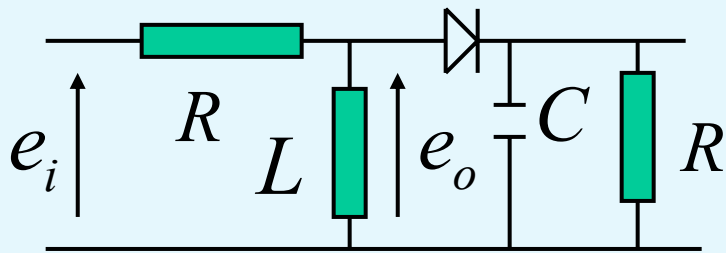


$$[10\omega_m + 3\omega_m \cos \omega_m t] \sin[10\omega_m t + 3 \sin \omega_m t]$$

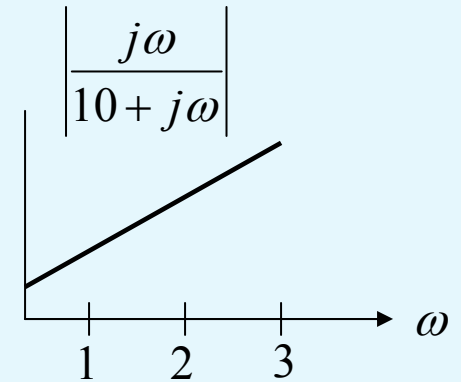
Demodulation of FM signals

- The differentiator can be approximated by any device whose magnitude transfer function is reasonably linear with the range of frequencies of interest.

Example



$$e_o(\omega) = \frac{j\omega L}{R + j\omega L} e_i(\omega)$$



$$\text{(Note: } F\left\{\frac{df(t)}{dt}\right\} = j\omega F(\omega) \text{)}$$